$\beta = \mu C_{ox} \frac{W}{L} = (350) \left( \frac{3.9 \cdot 8.85 \cdot 10^{-14}}{100 \cdot 10^{-8}} \right) \left( \frac{W}{L} \right) = 120 \frac{W}{L} \mu A / V^2$ 



2.2 In (a), the transistor sees  $V_{gs} = V_{DD}$  and  $V_{ds} = V_{DS}$ . The current is

$$I_{DS1} = \frac{\beta}{2} \left( V_{DD} - V_t - \frac{V_{DS}}{2} \right) V_{DS}$$

In (b), the bottom transistor sees  $V_{gs} = V_{DD}$  and  $V_{ds} = V_1$ . The top transistor sees  $V_{gs} = V_{DD} - V_1$  and  $V_{ds} = V_{DS} - V_1$ . The currents are

$$I_{DS2} = \beta \left( V_{DD} - V_t - \frac{V_1}{2} \right) V_1 = \beta \left( \left( V_{DD} - V_1 \right) - V_t - \frac{\left( V_{DS} - V_1 \right)}{2} \right) \left( V_{DS} - V_1 \right)$$

Solving for  $V_1$ , we find

$$V_{1} = (V_{DD} - V_{t}) - \sqrt{(V_{DD} - V_{t})^{2} - (V_{DD} - V_{t} - \frac{V_{DS}}{2})}V_{DS}$$

Substituting  $V_1$  indo the  $I_{DS2}$  equation and simplifying gives  $I_{DS1} = I_{DS2}$ .

- 2.3 The body effect does not change (a) because  $V_{sb} = 0$ . The body effect raises the threshold of the top transistor in (b) because  $V_{sb} > 0$ . This lowers the current through the series transistors, so  $I_{DS1} > I_{DS2}$ .
- 2.4  $C_{\text{permicron}} = \varepsilon L/t_{ox} = 3.9 * 8.85\text{e}{-}14 \text{ F/cm} * 90\text{e}{-}7 \text{ cm} / 16\text{e}{-}4 \text{ }\mu\text{m} = 1.94 \text{ fF/}\mu\text{m}.$

2.1

2.5 The minimum size diffusion contact is  $4 \times 5 \lambda$ , or  $1.2 \times 1.5 \mu m$ . The area is  $1.8 \mu m^2$  and perimeter is 5.4  $\mu m$ . Hence the total capacitance is

$$C_{db}(0V) = (1.8)(0.42) + (5.4)(0.33) = 2.54 \text{fF}$$

At a drain voltage of VDD, the capacitance reduces to

$$C_{db}(5V) = (1.8)(0.42)\left(1 + \frac{5}{0.98}\right)^{-0.44} + (5.4)(0.33)\left(1 + \frac{5}{0.98}\right)^{-0.12} = 1.78 \text{fF}$$

- 2.6 Set the two parts of EQ (2.26) equal at  $V_{ds} = V_{dsat.}$  Assume that EQ (2.27) is true and substitute it into (2.26) for  $V_{dsat.}$ , then simplify.
- 2.7 The new threshold voltage is found as

$$\begin{split} \phi_s &= 2(0.026) \ln \frac{2 \bullet 10^{17}}{1.45 \bullet 10^{10}} = 0.85V \\ \gamma &= \frac{100 \bullet 10^{-8}}{3.9 \bullet 8.85 \bullet 10^{-14}} \sqrt{2(1.6 \bullet 10^{-19})(11.7 \bullet 8.85 \bullet 10^{-14})(2 \bullet 10^{17})} = 0.75V^{1/2} \\ V_t &= 0.7 + \gamma \left(\sqrt{\phi_s + 4} - \sqrt{\phi_s}\right) = 1.66V \end{split}$$

The threshold increases by 0.96 V.

- 2.8 No. Any number of transistors may be placed in series, although the delay increases with the square of the number of series transistors.
- 2.9 The threshold is increased by applying a negative body voltage so  $V_{sb} > 0$ .

2.10 (a) 
$$(1.2 - 0.3)^2 / (1.2 - 0.4)^2 = 1.26 (26\%)$$

(b) 
$$\frac{e^{\frac{-0.3}{1.4 \cdot 0.026}}}{e^{\frac{-0.4}{1.4 \cdot 0.026}}} = 15.6$$
  
(c)  $v_T = kT/q = 34$  mV;  $\frac{e^{\frac{-0.3}{1.4 \cdot 0.034}}}{e^{\frac{-0.4}{1.4 \cdot 0.034}}} = 8.2$ ; note, however, that the total leakage

will normally be higher for both threshold voltages at high temperature.

2.11 The nMOS will be OFF and will see  $V_{ds} = V_{DD}$ , so its leakage is

$$I_{leak} = I_{dsn} = \beta v_T^2 e^{1.8} e^{\frac{-V_t}{m_T}} = 69 \, pA$$

2.12 If the voltage at the intermediate node is *x*, by KCL:

$$I_{leak} = \beta v_T^2 e^{1.8} e^{\frac{-V_t}{nv_T}} \left( 1 - e^{\frac{-x}{v_T}} \right) = \beta v_T^2 e^{1.8} e^{\frac{-(x+V_t)}{nv_T}}$$

Now, solve for x using n = 1:

$$\left(1 - e^{\frac{-x}{v_T}}\right) = e^{\frac{-x}{nv_T}} \to x = v_T \ln 2$$

Substituting, the current is exactly half that of the inverter.

2.13 Assume  $V_{DD} = 1.8$  V. For a single transistor with n = 1.4,

$$I_{leak} = I_{dsn} = \beta v_T^2 e^{1.8} e^{\frac{-V_t + \eta V_{DD}}{nv_T}} = 499 \, pA$$

For two transistors in series, the intermediate voltage *x* and leakage current are found as:

$$I_{leak} = \beta v_T^2 e^{1.8} e^{\frac{-V_t + \eta x}{nv_T}} \left( 1 - e^{\frac{-x}{v_T}} \right) = \beta v_T^2 e^{1.8} e^{\frac{\eta (V_{DD} - x) - V_t - x}{nv_T}}$$
$$e^{\frac{-V_t + \eta x}{nv_T}} \left( 1 - e^{\frac{-x}{v_T}} \right) = e^{\frac{\eta (V_{DD} - x) - V_t - x}{nv_T}}$$
$$x = 69 \text{ mV}; I_{leak} = 69 \text{ pA}$$

In summary, accounting for DIBL leads to more overall leakage in both cases. However, the leakage through series transistors is much less than half of that through a single transistor because the bottom transistor sees a small Vds and much less DIBL. This is called the *stack effect*.

For n = 1.0, the leakage currents through a single transistor and pair of transistors are 13.5 pA and 0.9 pA, respectively.